

## 6.300: Signal Processing

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### Discrete Fourier Transform (DFT)

**Analysis:** 
$$X[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-jk \frac{2\pi}{N} n}$$

**Synthesis:** 
$$x[n] = \sum_{k=0}^{N-1} X[k] e^{jk \frac{2\pi}{N} n}$$

#### Relating the DFT to DTFS and DTFT:

- DTFS of periodically-extended  $x_w[n] = x[n]w[n]$
- Sampled ( $\Omega \rightarrow 2\pi k/N$ ), scaled ( $1/N$ ) DTFT of  $x_w[n]$

**Frequency resolution:**  $\Delta f = f_s/N$  and  $\Delta\Omega = 2\pi/N$

# Agenda for Recitation

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- Fourier representations (so far)
- Spectral analysis with the DFT

What questions do you have from lecture?

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# Fourier Representations

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	type of time-domain signal			domain	period
CTFS	CT	period $T$	infinite length	integer $k$	none
DTFS	DT	period $N$	infinite length	integer $k$	$N$
CTFT	CT	aperiodic	infinite length	real $\omega$	none
DTFT	DT	aperiodic	infinite length	real $\Omega$	$2\pi$
DFT	DT	aperiodic	length $N$	integer $k$	$N$

The **discrete Fourier transform (DFT)** is a discrete-time, discrete-frequency Fourier transform.

- for aperiodic discrete-time ( $n$ ) signals  $x[n]$
- yields discrete-frequency ( $k$ ) representation  $X[k]$
- finite length ( $N$ ) in both time and frequency
- closely related to DTFS and DTFT

# Discrete-Time Fourier Representations

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Another Fourier transform? Why?

# DFT: Periodic Extension

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**Relation to DTFS:** The DFT yields the DTFS coefficients for a periodically-extended  $x_w[n] = x[n]w[n]$ .

$$\text{DTFS: } X[k] = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-jk \frac{2\pi}{N} n} \quad \text{DFT: } X[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-jk \frac{2\pi}{N} n}$$

Let  $N = 32$ . How many of the following signals have “simple” DFTs (i.e., mostly zeros)?

- $x_1[n] = \cos\left(\frac{2\pi}{32} n\right)$
- $x_2[n] = \cos\left(\frac{2\pi}{64} n\right)$
- $x_3[n] = \cos\left(\frac{2\pi}{16} n - \frac{\pi}{2}\right)$
- $x_4[n] = \cos\left(\frac{2\pi}{4} n\right) \cos\left(\frac{2\pi}{8} n\right)$
- $x_5[n] = \cos\left(\frac{2\pi}{32} n\right) + j \sin\left(\frac{2\pi}{32} n\right)$

# DFT: Sampling the DTFT

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**Relation to DTFT:** Multiplication by a window function  $w[n]$  in time corresponds to convolution with the Fourier transform of the window function,  $W(\Omega)$ , in frequency.

$$x_w[n] = x[n]w[n] \iff X_w(\Omega) = \frac{1}{2\pi}(X * W)(\Omega)$$

The DFT is a sampled ( $\Omega \rightarrow \frac{2\pi}{N}k$ ), scaled ( $1/N$ ) DTFT.

$$X[k] = \frac{1}{N}X_w\left(\frac{2\pi}{N}k\right)$$

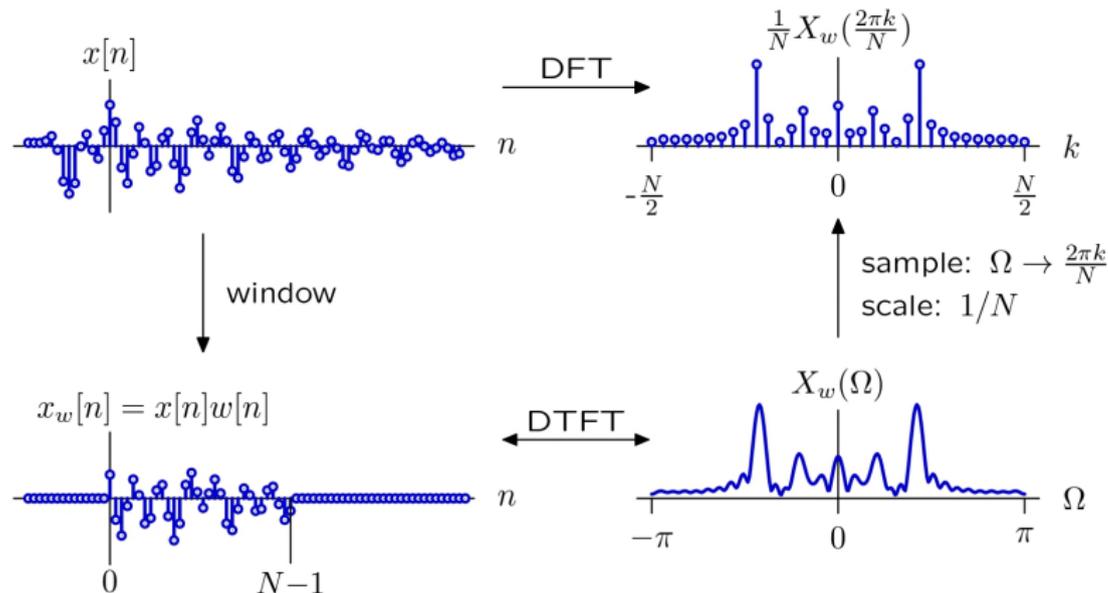
Notice that the distance between adjacent DFT frequency samples (or “bins”) is  $\Delta\Omega = 2\pi/N$  radians per sample. Accordingly, as  $N$  increases,  $\Delta\Omega$  decreases, and this yields finer frequency resolution.<sup>1</sup>

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<sup>1</sup>In general, the length of the DFT analysis window ( $N$ ) can exceed the length of the data. Ignore this case (“zero-padding”) for now.

## Relation Between DFT and DTFT

Graphical depiction of relation between DFT and DTFT.



Graphic: Professor Denny Freeman (freeman@mit.edu)

# Agenda for Recitation

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# Spectral Analysis with the DFT

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Consider the four discrete-time signals below.

$$x_1[n] = \cos\left(\frac{8\pi}{100}n\right)$$

$$x_2[n] = \cos\left(\frac{8\pi}{100}n - \frac{\pi}{4}\right)$$

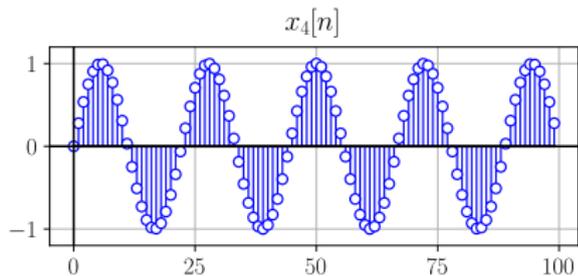
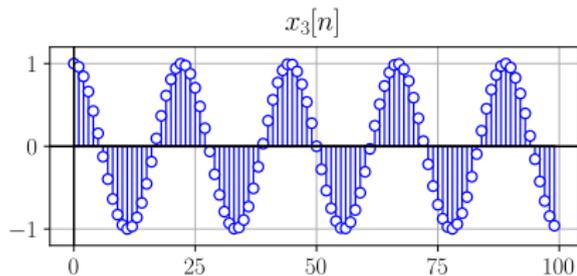
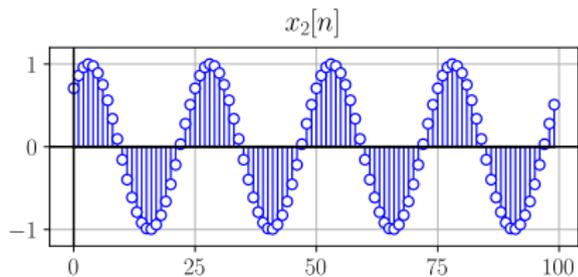
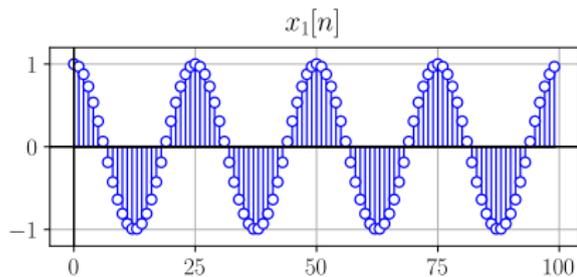
$$x_3[n] = \cos\left(\frac{9\pi}{100}n\right)$$

$$x_4[n] = \cos\left(\frac{9\pi}{100}n - \frac{\pi}{2}\right)$$

Suppose that the sampling rate is  $f_s = 44100$  samples per second. How would you write a Python program to generate a second's worth of samples for each of the signals above?

# Spectral Analysis with the DFT

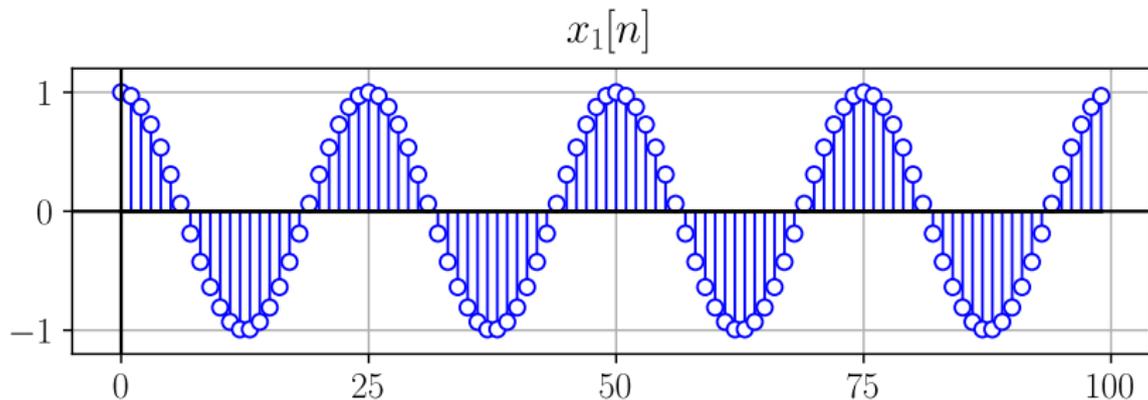
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# Spectral Analysis with the DFT

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The first  $N = 100$  samples of  $x_1[n]$  are plotted below.

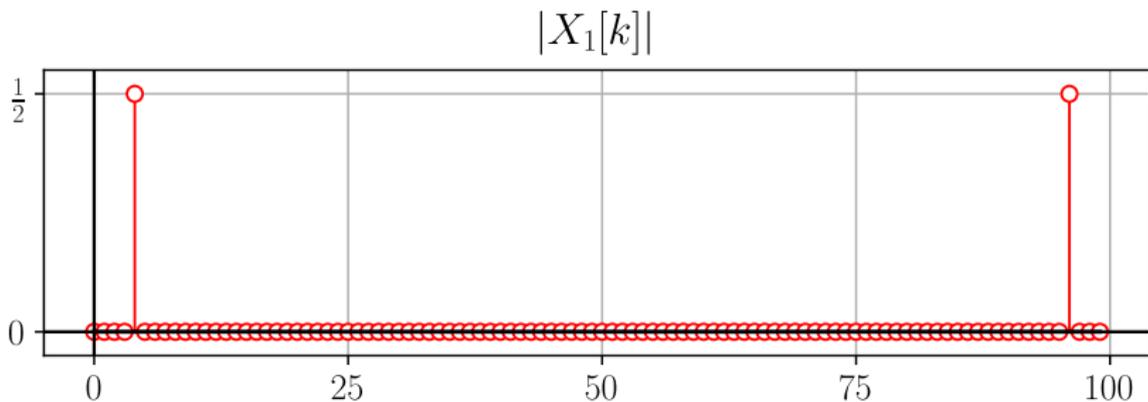


Recall that the sampling rate is  $f_s = 44100$  samples per second. What continuous-time cyclical frequency  $f_0$  (cycles per second) does discrete-time angular frequency  $\Omega_0 = \frac{8\pi}{100}$  (radians per sample) correspond to?

# Spectral Analysis with the DFT

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Shown below is  $|X_1[k]|$ , the magnitude of the DFT of the first  $N = 100$  samples of  $x_1[n]$ .

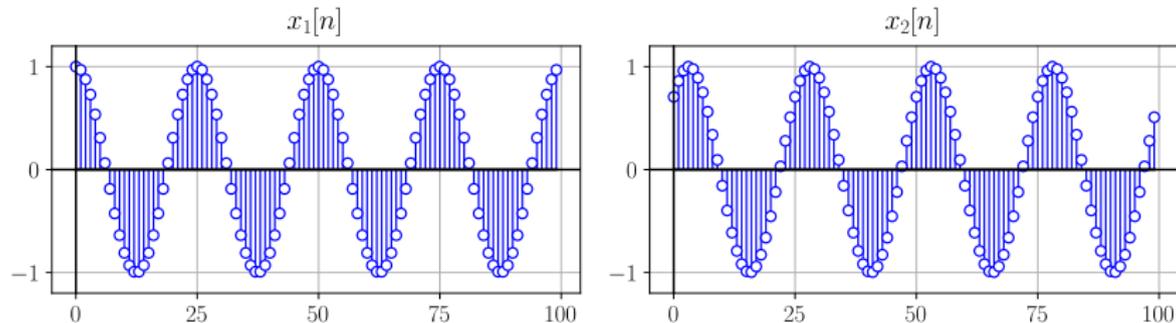


For which values of  $k$  is  $|X_1[k]|$  non-zero? Explain.

# Spectral Analysis with the DFT

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The first  $N = 100$  samples of  $x_1[n]$  and  $x_2[n]$  are plotted below.

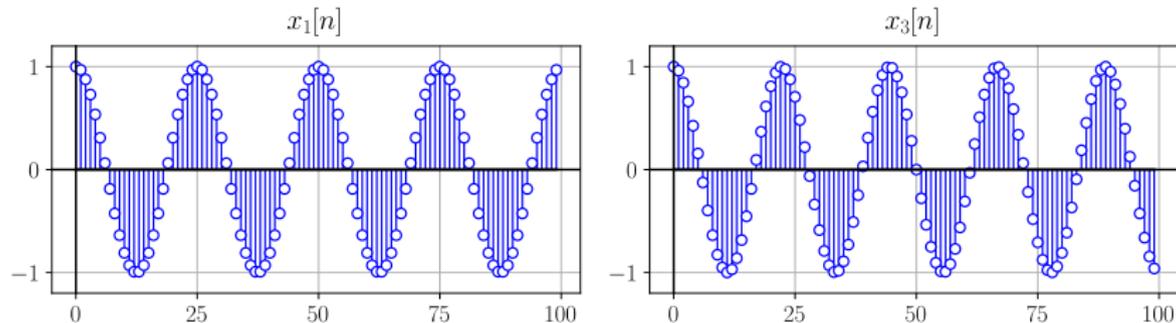


How will  $|X_1[k]|$  and  $|X_2[k]|$  differ?

# Spectral Analysis with the DFT

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The first  $N = 100$  samples of  $x_1[n]$  and  $x_3[n]$  are plotted below.

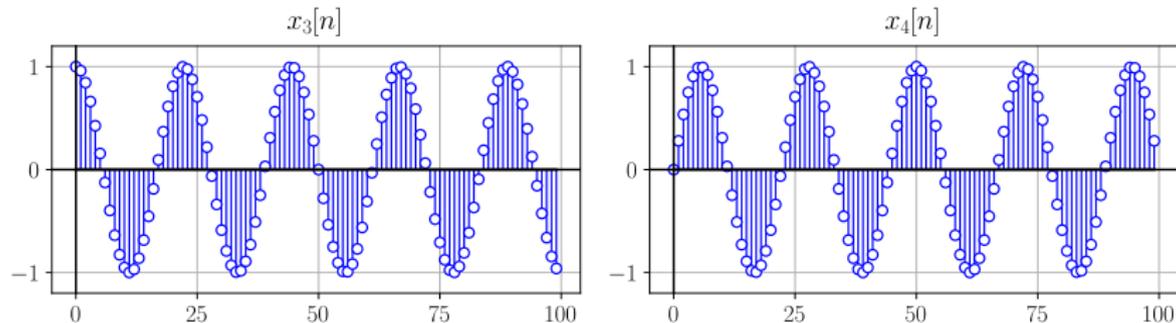


How will  $|X_1[k]|$  and  $|X_3[k]|$  differ?

# Spectral Analysis with the DFT

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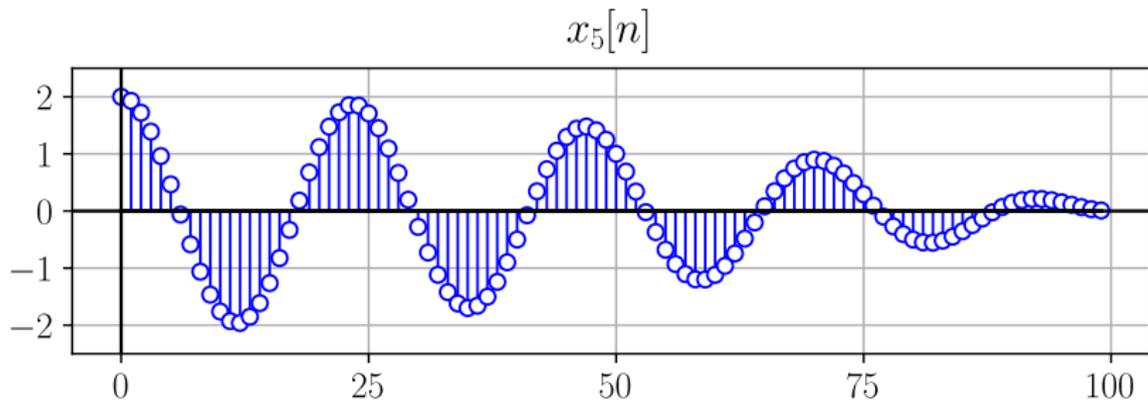
The first  $N = 100$  samples of  $x_3[n]$  and  $x_4[n]$  are plotted below.



How will  $|X_3[k]|$  and  $|X_4[k]|$  differ?

# Spectral Analysis with the DFT

Consider  $x_5[n] = \cos\left(\frac{8\pi}{100}n\right) + \cos\left(\frac{9\pi}{100}n\right)$ . The first  $N = 100$  samples are plotted below.



What is the minimum analysis window length  $N$  that's needed to resolve  $\Omega_1 = \frac{8\pi}{100}$  from  $\Omega_2 = \frac{9\pi}{100}$ ?

## Spectral Analysis with the DFT

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Frequency resolution is proportional to  $1/N$ . Suppose that we append a bunch of zeros to the end of our finite-length signal to make it longer. This increases  $N$ . Does our frequency resolution improve? Explain.

# Lessons Learned

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