

6.003 (Spring 2021)

*24 March 2021*

**Quiz 1: Solutions**

## 1. Sinusoids

### Part a

This cosine looks like it goes through one period around every 10 samples, so  $\Omega_1 \approx \frac{2\pi}{10}$ .

The amplitude looks like 2, but we start at a negative peak; so we could either say  $A_1 \approx 2$  and  $\phi_1 \approx \pm\pi$  (or any  $m\pi$ , where  $m$  is an odd integer); or we could say  $A_1 \approx -2$  and  $\phi_1 \approx 0$  (or any integer multiple of  $2\pi$ ).

### Part b

Here we have  $x_2[n] = a_2 e^{j\Omega_2 n} + a_2^* e^{-j\Omega_2 n} = 2|a_2| \cos(\Omega_2 n + \angle a_2)$ .

It looks like the amplitude of this wave is about 1, so  $|a_2| \approx 1/2$ .

The cosine seems to go through one full cycle every 15 samples, so  $\Omega_2 \approx 2\pi/15$ .

It also looks like it has been shifted to the right by 3/4 of a full cycle (or, equivalently, to the left by 1/4 of a full cycle), so  $\angle a_2 \approx 2\pi/4 = \pi/2$ .

### Part c

We can express  $x_3[n] = c_3 \cos \Omega_3 n + d_3 \sin \Omega_3 n$  as

$$\begin{aligned} x_3[n] &= c_3 \left( \frac{e^{j\Omega_3 n} + e^{-j\Omega_3 n}}{2} \right) + d_3 \left( \frac{e^{j\Omega_3 n} - e^{-j\Omega_3 n}}{2j} \right) \\ &= \frac{1}{2}(c_3 - jd_3)e^{j\Omega_3 n} + \frac{1}{2}(c_3 + jd_3)e^{-j\Omega_3 n} \\ &= \operatorname{Re} \left( (c_3 - jd_3)e^{j\Omega_3 n} \right) \end{aligned}$$

Since the period of  $x_3[n]$  is around 20, it follows that  $\Omega_3 \approx \frac{2\pi}{20}$

The peak amplitude of  $x_3[n]$  is approximately 1, so  $|c_3 - jd_3| = \sqrt{c_3^2 + d_3^2} = 1$ .

The first peak of  $x_3[n]$  occurs at about 1/8 of a cycle. So the angle of  $c_3 - jd_3$  must be approximately  $-\pi/4$ , meaning  $c_3 \approx d_3$ . Therefore,

$$c_3 \approx d_3 \approx \frac{1}{\sqrt{2}}$$

## 2. Fourier Transforms

### Part 1

Note that  $b = X(\omega = 0)$ , or, said another way,  $b$  is the DC component of  $x(\cdot)$ , which will be equal to the total "area under the curve" of  $x(\cdot)$ , so  $b = (t_2 - t_1)a$ .

### Part 2

Here, we probably want to solve for  $X(\omega)$  more generally, to get a sense of how the function behaves.

It will be simpler to consider a related function  $x_c(\cdot)$  that is a rectangular pulse of the same width and height, but centered around  $t = 0$  rather than being offset. In that case, since we have  $x(t) = x_c(t - (t_1 + t_2/2))$ , we know that:

$$X(\omega) = e^{-j\left(\frac{t_1+t_2}{2}\omega\right)} X_c(\omega).$$

We can then solve for  $X_c(\omega)$ :

$$\begin{aligned} X_c(\omega) &= \int_{-\infty}^{\infty} x_c(t) e^{-j\omega t} dt = \int_{-(t_2-t_1)/2}^{(t_2-t_1)/2} a e^{-j\omega t} dt = a \int_{-(t_2-t_1)/2}^{(t_2-t_1)/2} e^{-j\omega t} dt \\ &= -\frac{a}{j\omega} e^{-j\omega t} \Big|_{t=-(t_2-t_1)/2}^{(t_2-t_1)/2} = -\frac{a}{j\omega} \left( e^{-j\omega(t_2-t_1)/2} - e^{j\omega(t_2-t_1)/2} \right) = \frac{2a \sin\left(\frac{t_2-t_1}{2}\omega\right)}{\omega} \end{aligned}$$

So, all things considered, we have:

$$X(\omega) = e^{-j\left(\frac{t_1+t_2}{2}\omega\right)} X_c(\omega) = e^{-j\left(\frac{t_1+t_2}{2}\omega\right)} \frac{2a \sin\left(\frac{t_2-t_1}{2}\omega\right)}{\omega}$$

From this, we can see that  $|X(\Omega)| = |X_c(\Omega)| = \left| \frac{2a \sin\left(\frac{t_2-t_1}{2}\omega\right)}{\omega} \right|$

The value  $\omega_1$  is where this function first crosses zero, which will happen when the argument to sin is  $\pi$ , so we have:

$$\left( \frac{t_2 - t_1}{2} \right) \omega_1 = \pi \quad \Rightarrow \quad \omega_1 = \frac{2\pi}{t_2 - t_1}$$

### Parts 3 and 4

From the work above, we also know that  $\angle \{X(\Omega)\} = \angle \{X_c(\Omega)\} - \left(\frac{t_1+t_2}{2}\omega\right)$

So we should see a linear shape in the graph, with slope  $-\frac{t_1+t_2}{2}$ . However, we should also note that since  $X_c(\Omega)$  switched signs every  $\omega_1$  radians,  $\angle \{X_c(\Omega)\}$  was increasing (or decreasing) by  $\pi$  every  $\omega_1$  radians as well. So we not only to see a linear slope, but we also expect jumps upward every so often. The only graph that fits the bill is graph **F**.

The value  $c$  corresponds to the slope, which should be  $-\frac{t_1+t_2}{2}$ , as seen above.

### 3. Related Transforms

#### Part 1

$$f_2[n] = -f[-n] \quad \Rightarrow \quad F_2(\Omega) = -F(-\Omega)$$

#### Part 2

$$f_3[n] = f[n] - \delta[n] \quad \Rightarrow \quad F_3(\Omega) = F(\Omega) - 1$$

#### Part 3

$$f_4[n] = \text{Asym} \{f[n]\} = \frac{f[n] - f[-n]}{2} \quad \Rightarrow \quad F_4(\Omega) = j\text{Im} \{F(\Omega)\}$$

#### Part 4

$$f_5[n] = f[n+2] \quad \Rightarrow \quad F_5(\Omega) = e^{j2\Omega} F(\Omega)$$

#### Part 5

If we let:

$$f_s[n] = \begin{cases} f[n/3] & \text{if } n \text{ is divisible by } 3 \\ 0 & \text{otherwise} \end{cases}$$

Then we have  $f_6[n] = f_s[n] + f_s[n-1] + f_s[n+1]$ , so:

$$F_6(\Omega) = F_s(\Omega) (1 + e^{j\Omega} + e^{-j\Omega}) = F_s(\Omega) (1 + 2 \cos(\Omega))$$

Since  $F_s(\Omega) = F(3\Omega)$ , we have:

$$F_6(\Omega) = (F(3\Omega)) (1 + 2 \cos(\Omega))$$

## 4. DFT

### Part 1

We know that  $k = f \frac{N}{f_s}$ , so since  $k = \pm 268 \approx \pm \frac{f}{2}$ , we know that we must have  $N \approx \frac{f_s}{2}$ , so the only real possibility is  $N = 1024$ .

### Part 2

Importantly, since the "A" note ( $f = 1760\text{Hz}$ ) is above the Nyquist rate, it is going to alias down to  $f_{\text{apparent}} = f_s - 1760\text{Hz} = 240\text{Hz}$ . Converting to  $k$ , we find  $k = \pm 240 \left(\frac{1024}{2000}\right) \approx \pm 123$ .

We could also find this by first finding the expected  $k$  value for  $f = 1760\text{Hz}$  first:

$k = \pm \left(\frac{1024}{2000}\right) 1760 \approx \pm 901$ . But that's not in the range  $[0, N/2]$ , so it can't be the final answer. Since the coefficients are periodic in  $N$ , we need to find a value  $1024m + 901$ , where  $m$  is an integer, for which that value falls in the range  $[0, 512]$ . That happens to be  $1024 - 901 = 123$ .

### Part 3

Since the "C" note shows up at  $k = \pm 268$  and the "A" note (due to aliasing) shows up at  $k = \pm 123$ , the "A" note must correspond to the peaks that are closer to 0, i.e.,  $k_1$ .

## 5. DTFS Components

The first six of these problems can be solved directly:

### Part 0

$$X_0[k] = 1/6$$

This signal has a constant magnitude of  $1/6$  (graph **G**), a constant real part of  $1/6$  (graph **G**), and a constant imaginary part of  $0$  (graph **C**).

Looking ahead, importantly, since all of  $x_0$ ,  $x_1$ ,  $x_2$ ,  $x_3$ ,  $x_4$ , and  $x_5$  are just time-shifted versions of each other, they should all have the same magnitude (graph **G**).

### Part 1

$$X_1[k] = \frac{1}{6}e^{-j\frac{2\pi}{6}k} = \frac{1}{6} \left( \cos\left(\frac{2\pi}{6}k\right) - j \sin\left(\frac{2\pi}{6}k\right) \right)$$

The magnitude graph is still **G**, but now the real part looks like a cosine that goes through one period in  $K = 6$  (graph **I**), and the imaginary part looks like an inverted sine that goes through one period in  $K = 6$  (graph **D**).

### Part 2

$$X_2[k] = \frac{1}{6}e^{-j\frac{4\pi}{6}k} = \frac{1}{6} \left( \cos\left(\frac{4\pi}{6}k\right) - j \sin\left(\frac{4\pi}{6}k\right) \right)$$

This is the same, but now the cosine and sine each go through two full periods in  $K = 6$ , so magnitude is still **G**, real part is **K**, and imaginary part is **E**.

### Part 3

This one is interesting, in that  $X_3[k]$  ends up being purely real:

$$X_3[k] = \frac{1}{6}e^{-j\pi k} = \frac{1}{6}(-1)^k$$

Magnitude is **G**, real part is **M**, and imaginary part is **C** (no imaginary part).

### Part 4

$$X_4[k] = \frac{1}{6}e^{-j\frac{8\pi}{6}k} = \frac{1}{6}e^{j\frac{4\pi}{6}k} = \frac{1}{6} \left( \cos\left(\frac{4\pi}{6}k\right) + j \sin\left(\frac{4\pi}{6}k\right) \right)$$

This looks like  $X_2[k]$ , but with the imaginary part negated, so magnitude is **G**, real part is **K**, and imaginary part is **A**.

### Part 5

$$X_5[k] = \frac{1}{6}e^{-j\frac{10\pi}{6}k} = \frac{1}{6}e^{j\frac{2\pi}{6}k} = \frac{1}{6} \left( \cos\left(\frac{2\pi}{6}k\right) + j \sin\left(\frac{2\pi}{6}k\right) \right)$$

This looks like  $X_1[k]$ , but with the imaginary part negated, so magnitude is **G**, real part is **I**, and imaginary part is **B**.

The next 6 parts all have simple relationships to the first 6 parts. In particular, note that for all  $i$ ,  $y_i[n] = 1 - x_i[n]$ . Thus, by linearity, we have  $Y_i[k] = \delta[k] - X_i[k]$ , which we can use to simplify our analysis.

Since all of the signals  $y_i$  are just time-shifted versions of each other, all of the  $Y_i$  will have the same magnitude.

And for each  $Y_i[k]$ , we will have  $\text{Re}(Y_i[0]) = 1 - \text{Re}(X_i[0])$ , and for all other  $k$ ,  $\text{Re}(Y_i[k]) = -\text{Re}(X_i[k])$ .

Similarly, for all  $k$ ,  $\text{Im}(Y_i[k]) = -\text{Im}(X_i[k])$ .

### Part 6

$$Y_0[k] = \delta[k] - X_0[k] = \delta[k] = 1/6$$

Magnitude **O**, real part **T**, imaginary part **C**.

### Part 7

$$Y_1[k] = \delta[k] - X_1[k]$$

Magnitude **O**, real part **S**, imaginary part **B**.

### Part 8

$$Y_2[k] = \delta[k] - X_2[k]$$

Magnitude **O**, real part **Q**, imaginary part **A**.

### Part 9

$$Y_3[k] = \delta[k] - X_3[k]$$

Magnitude **O**, real part **N**, imaginary part **C**.

### Part 10

$$Y_4[k] = \delta[k] - X_4[k]$$

Magnitude **O**, real part **Q**, imaginary part **E**.

### Part 10

$$Y_5[k] = \delta[k] - X_5[k]$$

Magnitude **O**, real part **S**, imaginary part **D**.