## 6. 300 Signal Processing

# Week 4, Lecture A: Continuous Time Fourier Transform

- Definition
- Example
- Impulse function δ(t)

#### Quiz 1: Thursday October 3, 2-4pm 50-340

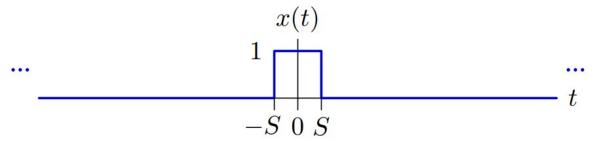
- Closed book except for one page of notes (8.5" x 11" both sides)
- No electronic devices (No headphones, cell phones, calculators, ...)
- Coverage up to Week #3 (DTFS)

#### From Periodic to Aperiodic

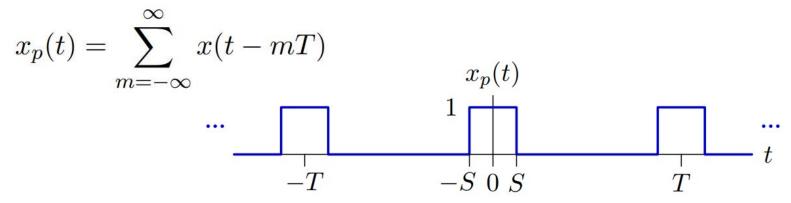
- Previously, we have focused on Fourier representations of periodic signals: e.g., sounds, waves, music, ...
- However, most real-world signals are not periodic.

Today: generalizing Fourier representations to include aperiodic signals -> Fourier Transform

How can we represent an aperiodic signal as a sum of sinusoids?



Strategy: make a periodic version of x(t) by summing shifted copies:



Since  $x_p(t)$  is periodic, it has a Fourier series (which depends on T) Find Fourier series coefficients  $X_p[k]$  and take the limit of  $X_p[k]$  as T  $\rightarrow \infty$ As T  $\rightarrow \infty$ ,  $x_p(t) \rightarrow x(t)$  and Fourier series will approach Fourier transform.

$$x_p(t) = \sum_{m = -\infty} x(t - mT)$$

$$\vdots$$

$$\vdots$$

$$T$$

$$\vdots$$

$$\vdots$$

$$T$$

$$\vdots$$

$$T$$

$$\vdots$$

$$T$$

Calculate the Fourier series coefficients  $X_p[k]: X_p[k] = \frac{1}{T} \int_{-T/2}^{T/2} x_p(t) \cdot e^{-j\frac{2\pi}{T}kt} dt$ 

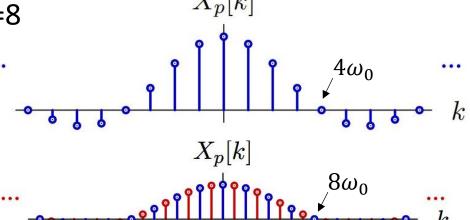
$$X_{p}[k] = \frac{1}{T} \int_{-T/2}^{T/2} x(t) \cdot e^{-j\frac{2\pi}{T}kt} dt = \frac{1}{T} \int_{-S}^{S} 1 \cdot e^{-j\frac{2\pi}{T}kt} dt = \frac{1}{T} \cdot \frac{e^{-j\frac{2\pi}{T}kt}}{(-j2\pi k/T)} \bigg|_{-S}^{S} = \frac{2\sin(\frac{2\pi k}{T}S)}{T(\frac{2\pi k}{T})}$$

Plot the resulting Fourier coefficients when S=1 and T=8

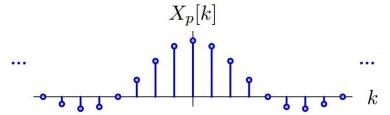
What happens if you double the period T?

There are twice as many samples per period of the sine function

The red samples are at new intermediate frequencies



$$X_p[k] = \frac{2\sin(\frac{2\pi k}{T}S)}{T(\frac{2\pi k}{T})} \qquad \dots$$



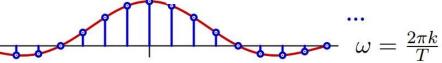
let  $\omega = \frac{2\pi k}{T}$ , Define a new function  $X(\omega) = T \cdot X_p[k] = 2\frac{\sin(\omega S)}{\omega}$ 

If we consider  $\omega$  and  $X(\omega) = 2\frac{\sin(\omega S)}{\omega}$  to be continuous,  $TX_p[k]$  represents a sampled version of the

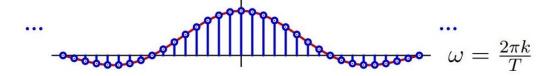
function  $X(\omega)$ .

$$TX_p[k] = X\left(\omega = \frac{2\pi k}{T}\right)$$

$$S=1$$
 and  $T=8$ :

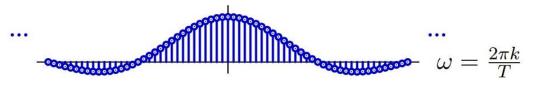


$$S=1$$
 and  $T=16$ :



As T increases, the resolution in  $\omega$  increases.

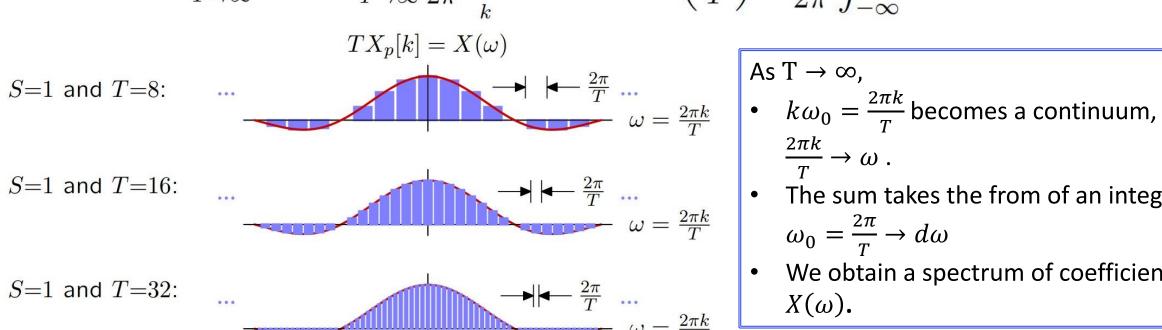
$$S=1$$
 and  $T=32$ :



We can reconstruct x(t) from  $X(\omega)$  using Riemann sums (approximating an integral by a finite sum).

$$x_p(t) = \sum_{k=-\infty}^{\infty} X_p[k] e^{j\frac{2\pi}{T}kt} = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} T X_p[k] e^{j\frac{2\pi}{T}kt} \left(\frac{2\pi}{T}\right)$$

$$x(t) = \lim_{T \to \infty} x_p(t) = \lim_{T \to \infty} \frac{1}{2\pi} \sum_{k} TX_p[k] e^{j\frac{2\pi}{T}kt} \left(\frac{2\pi}{T}\right) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$



- $\frac{2\pi k}{T} \to \omega$ .
- The sum takes the from of an integral,  $\omega_0 = \frac{2\pi}{T} \to d\omega$
- We obtain a spectrum of coefficients:  $X(\omega)$ .

#### **Fourier Transform**

$$x(t) = \lim_{T \to \infty} x_p(t) = \lim_{T \to \infty} \frac{1}{2\pi} \sum_{k} T X_p[k] e^{j\frac{2\pi}{T}kt} \left(\frac{2\pi}{T}\right) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

Since 
$$X(\omega) = T \cdot X_p[k]$$
 
$$X_p[k] = \frac{1}{T} \int_{-T/2}^{T/2} x(t) \cdot e^{-j\frac{2\pi}{T}kt} dt$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) \cdot e^{-j\omega t} dt$$

Fourier series and transforms are similar: both represent signals by their frequency content.

#### **Continuous-Time Fourier Transform**

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) \cdot e^{j\omega t} d\omega$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) \cdot e^{-j\omega t} dt$$

Synthesis equation

**Analysis equation** 

#### **Continuous-Time Fourier Series**

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} X[k]e^{j\frac{2\pi kt}{T}}$$

$$X[k] = \frac{1}{T} \int_{T} x(t)e^{-j\frac{2\pi}{T}kt}dt$$

**Synthesis equation** 

$$\omega_0 = \frac{2\pi}{T}$$

**Analysis equation** 

Periodic signals can be synthesized from a discrete set of harmonics. Aperiodic signals generally require all possible frequencies.

#### **Continuous-Time Fourier Transform**

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) \cdot e^{j\omega t} \, d\omega$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) \cdot e^{-j\omega t} dt$$

**Synthesis equation** 

**Analysis equation** 

#### **Continuous-Time Fourier Series**

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} X[k]e^{j\frac{2\pi kt}{T}}$$

$$X[k] = \frac{1}{T} \int_{T} x(t)e^{-j\frac{2\pi}{T}kt}dt$$

**Synthesis equation** 

$$\omega_0 = \frac{2\pi}{T}$$

**Analysis equation** 

All of the information in a periodic signal is contained in one period. The information in an aperiodic signal is spread across all time.

#### **Continuous-Time Fourier Transform**

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) \cdot e^{j\omega t} d\omega$$

**Synthesis equation** 

$$X(\omega) = \int_{-\infty}^{\infty} x(t) \cdot e^{-j\omega t} \, dt$$

**Analysis equation** 

#### **Continuous-Time Fourier Series**

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} X[k]e^{j\frac{2\pi kt}{T}}$$

$$\omega_0 = \frac{2\pi}{T}$$

$$X[k] = \frac{1}{T} \int_{T} x(t)e^{-j\frac{2\pi}{T}kt} dt$$

Harmonic frequencies  $k\omega_0$  are samples of continuous frequency  $\omega$ 

#### **Continuous-Time Fourier Transform**

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) \cdot e^{j\omega t} d\omega$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) \cdot e^{-j\omega t} dt$$

**Synthesis equation** 

**Analysis equation** 

#### **Continuous-Time Fourier Series**

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} X[k]e^{j\frac{2\pi k}{T}t}$$

$$X[k] = \frac{1}{T} \int_{T} x(t)e^{-j\frac{2\pi}{T}kt}dt$$

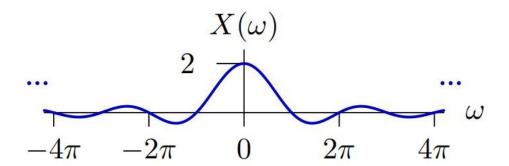
**Synthesis equation** 

$$\omega_0 = \frac{2\pi}{T}$$

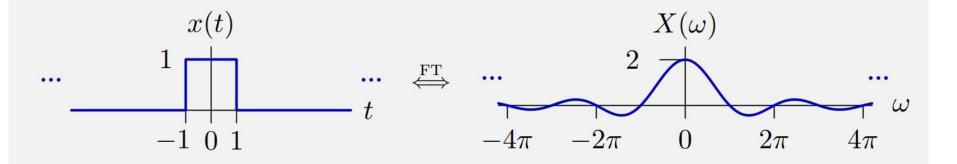
**Analysis equation** 

$$x(t) = \begin{cases} 1 & -1 < t < 1 \\ 0 & \text{otherwise} \end{cases} \dots \underbrace{ 1 \qquad 1 \qquad 1 \qquad \dots }_{-1,0,1}$$

$$X(\omega) = \int_{-\infty}^{\infty} x(t) \cdot e^{-j\omega t} dt = \int_{-1}^{1} 1 \cdot e^{-j\omega t} dt = \frac{e^{-j\omega t}}{-j\omega} \bigg|_{-1}^{1} = 2\frac{\sin(\omega)}{\omega}$$



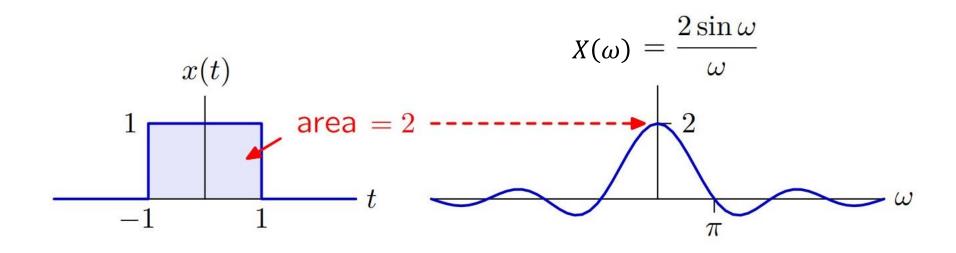
The Fourier transform of a rectangular pulse is  $2\frac{\sin\omega}{\omega}$ .



 $X(\omega)$  contains all frequencies  $\omega$  except non-zero multiples of  $\pi$ .

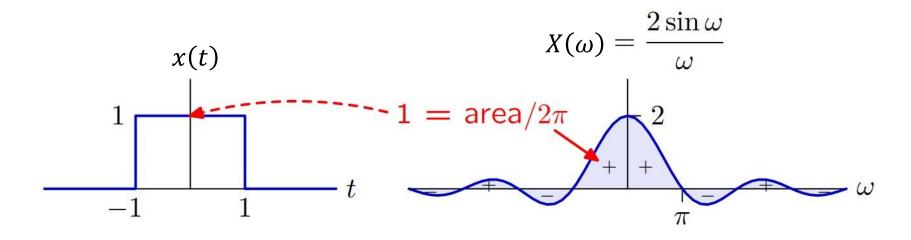
$$X(\omega = m\pi) = \int_{-1}^{1} e^{-j\omega t} dt = \int_{-1}^{1} e^{-jm\pi t} dt = \begin{cases} 2 & \text{if } m = 0 \\ 0 & \text{otherwise} \end{cases}$$

By definition, the value of  $X(\omega = 0)$  is the integral of x(t) over all time



$$X(0) = \int_{-\infty}^{\infty} x(t)e^{-j0t}dt = \int_{-\infty}^{\infty} x(t)dt$$

By definition, the value of x(t=0) is the integral of  $X(\omega)$  over all frequencies, divided by  $2\pi$ 



$$x(0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega 0} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) d\omega$$

### **Check yourself!**

Signal  $x_2(t)$  and its Fourier transform  $X_2(\omega)$  are shown below.



Which of the following is true?

1. 
$$b=2$$
 and  $\omega_0=\pi/2$ 

2. 
$$b=2$$
 and  $\omega_0=2\pi$ 

3. 
$$b=4$$
 and  $\omega_0=\pi/2$ 

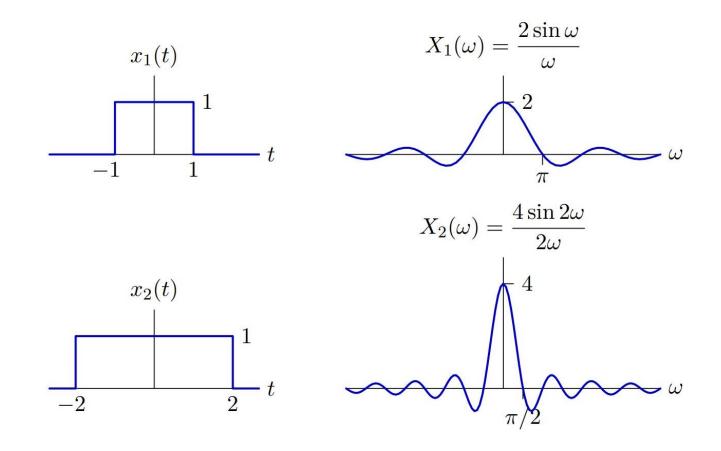
4. 
$$b=4$$
 and  $\omega_0=2\pi$ 

5. none of the above

$$X_2(\omega) = \int_{-\infty}^{\infty} x_2(t) \cdot e^{-j\omega t} dt = \int_{-2}^{2} 1 \cdot e^{-j\omega t} dt = \frac{e^{-j\omega t}}{-j\omega} \bigg|_{-2}^{2} = 2\frac{\sin(2\omega)}{\omega} = \frac{4\sin(2\omega)}{2\omega}$$

### **Stretching In Time**

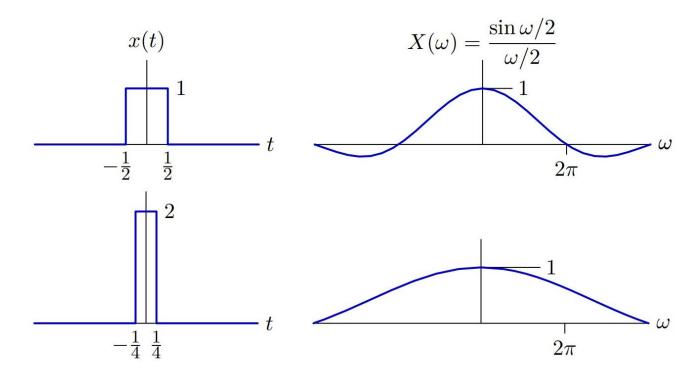
How would  $X(\omega)$  scale if time were stretched?



Stretching in time compresses in frequency.

#### **Compressing Time to the Limit**

Alternatively, compress time while keeping area = 1:



In the limit, the pulse has zero width but area 1! We represent this limit with the delta function:  $\delta(t)$ .



Although physically unrealizable, the impulse (a.k.a. Dirac delta) function  $\delta(t)$  is useful as a mathematically tractable approximation to a very brief signal.

 $\delta(t)$  only has a nonzero value at t = 0, but it has finite area: it is most easily described as an integral:

$$\int_{-\infty}^{\infty} \delta(t)dt = \int_{0_{-}}^{0_{+}} \delta(t)dt = 1 \qquad \qquad \int_{-\infty}^{\infty} \delta(t-a) \ dt = \int_{a_{-}}^{a_{+}} \delta(t) \ dt = 1$$

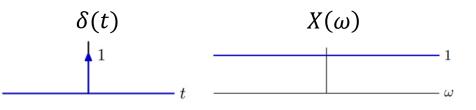
Importantly, it has the following property (the "sifting property"):

$$\int_{-\infty}^{\infty} \delta(t-a)f(t)dt = f(a)$$

let 
$$\tau = t - a$$
,  $\int_{-\infty}^{\infty} \delta(\tau) f(\tau + a) d\tau = \int_{0_{-}}^{0_{+}} \delta(\tau) f(a) d\tau = f(a) \cdot \int_{0_{-}}^{0_{+}} \delta(\tau) d\tau = f(a)$ 

The Fourier Transform of  $\delta(t)$ :

$$X(\omega) = \int_{-\infty}^{\infty} \delta(t) \cdot e^{-j\omega t} dt = \int_{0_{-}}^{0_{+}} \delta(t) \cdot e^{-j\omega 0} dt = 1$$



Find the function whose Fourier transform is a unit impulse.

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \delta(\omega) \cdot e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{0_{-}}^{0_{+}} \delta(\omega) \cdot e^{j0t} d\omega = \frac{1}{2\pi}$$

$$1 \ \stackrel{ ext{CTFT}}{\Longleftrightarrow} \ 2\pi\delta(\omega)$$

Notice the similarity to the previous result:

$$\delta(t) \ \stackrel{ ext{CTFT}}{\Longleftrightarrow} \ 1$$

These relations are duals of each other:

- A constant in time consists of a single frequency at  $\omega = 0$ .
- An impulse in time contains components at all frequencies.

Although physically unrealizable, the impulse (a.k.a. Dirac delta) function is useful as a mathematically tractable approximation to a very brief signal.

Find the function whose Fourier transform is a shifted impulse.

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} \delta(\omega - \omega_0) \cdot e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} \delta(\omega - \omega_0) \cdot e^{j\omega_0 t} d\omega$$
$$= \frac{1}{2\pi} e^{j\omega_0 t} \int_{-\infty}^{\infty} \delta(\omega - \omega_0) d\omega$$
$$= \frac{1}{2\pi} e^{j\omega_0 t}$$

$$e^{j\omega_o t} \stackrel{\text{CTFT}}{\Longrightarrow} 2\pi\delta(\omega-\omega_o)$$

We can use this result to relate Fourier series to Fourier Transforms.

If a periodic signal f(t) = f(t + T) has a Fourier Series representation, then it can also be represented by an equivalent Fourier Transform.

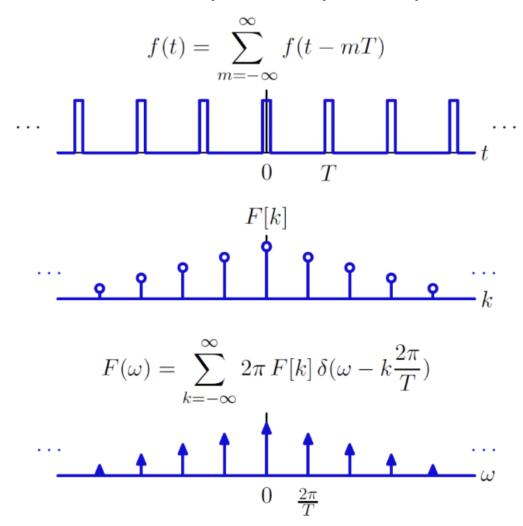
$$e^{j\omega_0 t} \overset{\text{FT}}{\Longrightarrow} 2\pi\delta(\omega - \omega_0)$$

$$f(t) = f(t+T) = \sum_{k=-\infty}^{\infty} F[k]e^{j\frac{2\pi}{T}kt} \quad \overset{\text{CTFS}}{\longleftrightarrow} \qquad F[k]$$

$$f(t) = f(t+T) = \sum_{k=-\infty}^{\infty} F[k]e^{j\frac{2\pi}{T}kt} \quad \overset{\text{CTFT}}{\longleftrightarrow} \qquad \sum_{k=-\infty}^{\infty} 2\pi F[k]\delta\left(\omega - \frac{2\pi}{T}k\right) = F(\omega)$$

Each term in the Fourier Series is replaced by an impulse in the Fourier transform.

Each Fourier Series term is replaced by an impulse in the Fourier transform.



### **Summary**

We will now go to 4-370 for recitation & common hour

• Continuous Time Fourier Transform: Fourier representation to all CT signals!

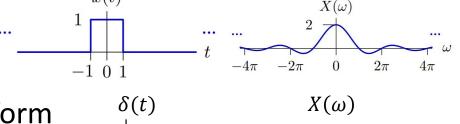
$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) \cdot e^{j\omega t} d\omega$$

**Synthesis equation** 

$$X(\omega) = \int_{-\infty}^{\infty} x(t) \cdot e^{-j\omega t} dt$$

**Analysis equation** 

- Very useful signals:
  - Rectangular pulse and its Fourier Transform (sinc)
  - Delta function (Unit impulse) and its Fourier Transform



• If a periodic signal f(t) = f(t + T) has a Fourier Series representation, then it can also be represented by an equivalent Fourier Transform.