

# 6.003: Signal Processing

## Convolution and Filtering

time domain

$$y(t) = (h * x)(t) = \int h(\tau)x(t - \tau) d\tau$$

$$y[n] = (h * x)[n] = \sum_m h[m]x[n - m]$$

frequency domain

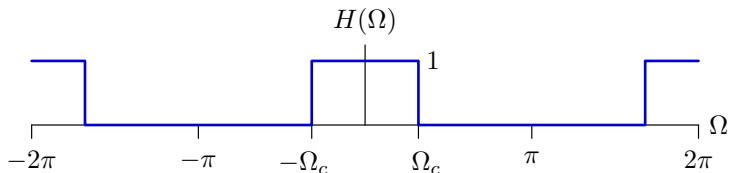
$$Y(\omega) = H(\omega)X(\omega)$$

$$Y(\Omega) = H(\Omega)X(\Omega)$$

## The “Ideal” Low-Pass Filter

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Consider a system characterized by the following purely real frequency response:

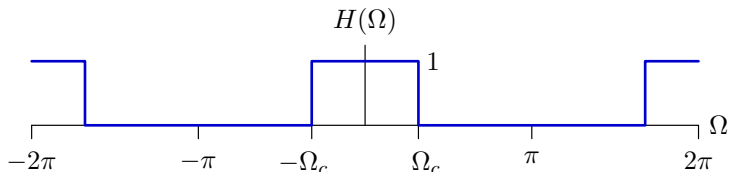


Such a system is called a **low-pass filter**, because it allows low frequencies to pass through unmodified, while attenuating high frequencies.

We could apply this filter to a signal by multiplying the DTFT of that signal by the values above. But we could also apply the filter by operating in the time domain.

## The “Ideal” Low-Pass Filter

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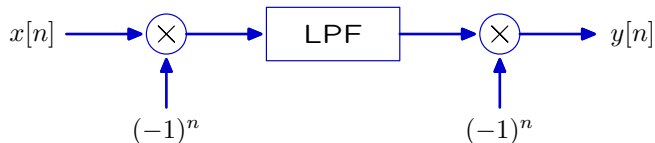
We can apply this filter to a signal by convolving with its unit sample response. What is the unit sample response of the system whose frequency response is shown above?

$$\begin{aligned}h[n] &= \frac{1}{2\pi} \int_{-2\pi}^{2\pi} H(\Omega) e^{j\Omega n} d\Omega \\&= \frac{1}{2\pi} \int_{-\Omega_c}^{\Omega_c} e^{j\Omega n} d\Omega \\&= \frac{1}{2\pi} \frac{1}{jn} e^{j\Omega n} \Big|_{\Omega=-\Omega_c}^{\Omega_c} \\&= \frac{\sin(\Omega_c n)}{\pi n}\end{aligned}$$

## Cascaded System

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Consider the following system, where LPF represents a lowpass filter of the form discussed on the previous slides.



How many of the following statements are true?

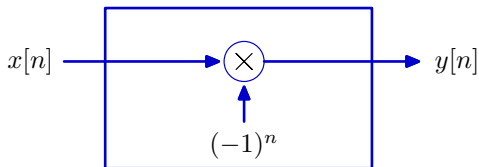
- The transformation from  $x[n]$  to  $y[n]$  is linear.
- The transformation from  $x[n]$  to  $y[n]$  is time invariant.
- The transformation from  $x[n]$  to  $y[n]$  is a high-pass filter.

## Consider Each Part Separately

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Start with the multiplier.

Is the following system linear?



Assume the response to the input  $x_1[n]$  is  $y_1[n]$  and the response to the input  $x_2[n]$  is  $y_2[n]$ . Calculate the response to  $x[n] = \alpha x_1[n] + \beta x_2[n]$ .

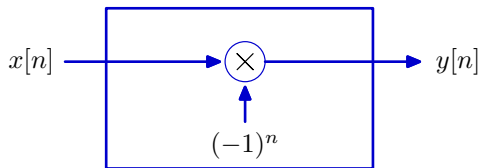
$$\begin{aligned}y[n] &= (-1)^n x[n] \\ &= (-1)^n (\alpha x_1[n] + \beta x_2[n]) \\ &= \alpha (-1)^n x_1[n] + \beta (-1)^n x_2[n] \\ &= \alpha y_1[n] + \beta y_2[n]\end{aligned}$$

$\therefore$  linear!

## Consider Each Part Separately

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Is this system time-invariant?



Assume the response to the input  $x_1[n]$  is  $y_1[n]$ .

Calculate the response to  $x[n] = x_1[n - n_0]$ .

$$\begin{aligned}y[n] &= (-1)^n x[n] \\ &= (-1)^n x_1[n - n_0] \\ &= (-1)^{n_0} (-1)^{n-n_0} x_1[n - n_0] \\ &= (-1)^{n_0} y_1[n - n_0]\end{aligned}$$

Shifting the input by  $n_0$  samples shifts does not just shift the output by  $n_0$  samples.

$\therefore$  this system is not time-invariant!

## Consider Each Part Separately

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Is the LPF linear? time-invariant?



The defining property of a **filter** is that its output is a weighted sum of (possibly) shifted versions of the frequencies in the input, so that

$$Y(\Omega) = H(\Omega)X(\Omega).$$

Multiplication in frequency is the same as convolution in time, so

$$y[n] = (x * h_L)[n]$$

where  $h_L[\cdot]$  represents the unit-sample response of the filter.

Convolution is both linear and time invariant (as shown on the next slides).

## Convolution is Linear

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Assume the response to the input  $x_1[n]$  is  $y_1[n]$  and the response to the input  $x_2[n]$  is  $y_2[n]$ . Calculate the response to  $x[n] = \alpha x_1[n] + \beta x_2[n]$ .

$$\begin{aligned}y[n] &= \sum_{m=-\infty}^{\infty} x[m]h[n-m] \\&= \sum_{m=-\infty}^{\infty} (\alpha x_1[m] + \beta x_2[m])h[n-m] \\&= \alpha \sum_{m=-\infty}^{\infty} x_1[m]h[n-m] + \beta \sum_{m=-\infty}^{\infty} x_2[m]h[n-m] \\&= \alpha y_1[n] + \beta y_2[n]\end{aligned}$$

$\therefore$  linear!



## Convolution is Time-Invariant

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Assume the response to the input  $x_1[n]$  is  $y_1[n] = (x_1 * h)[n]$ .

Calculate the response to  $x[n] = x_1[n - n_0]$ .

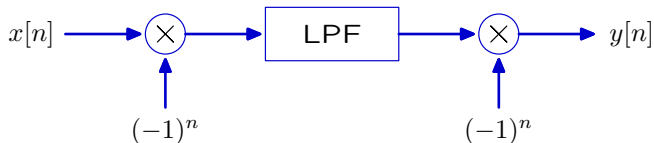
$$\begin{aligned}y[n] &= (x * h)[n] \\&= \sum_{m=-\infty}^{\infty} x[m]h[n-m] \\&= \sum_{m=-\infty}^{\infty} x_1[m-n_0]h[n-m] \\&= \sum_{l=-\infty}^{\infty} x_1[l]h[n-(n_0+l)] \\&= \sum_{l=-\infty}^{\infty} x_1[l]h[(n-n_0)-l] \\&= (x_1 * h)[n - n_0] = y_1[n - n_0]\end{aligned}$$

$\therefore$  time-invariant!

## Cascaded System

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This system is the cascade of three linear subsystems, two of which are time varying.



Thus we know that the composite system is linear.

- The transformation from  $x[n]$  to  $y[n]$  is linear. ✓
- The transformation from  $x[n]$  to  $y[n]$  is time invariant.
- The transformation from  $x[n]$  to  $y[n]$  is a high-pass filter.

The first is true. Not sure about the others.

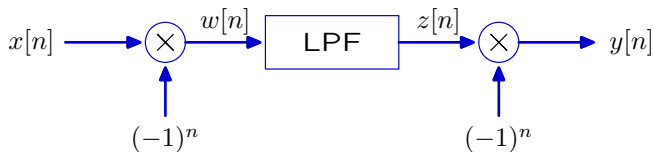
Could the cascade of two time-varying systems be time-invariant?

Yes. Think about the cascade of  $\times(-1)^n$  with  $\times(-1)^n$ .

## Cascaded System

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Determine an expression for  $y[n]$  in terms of  $x[n]$  using a time-domain approach. Assume that the unit sample response of the LPF is  $h_L[n]$ .



$$w[n] = (-1)^n x[n]$$

$$z[n] = (w * h_L)[n] = \sum_{m=-\infty}^{\infty} (-1)^m x[m] h_L[n-m]$$

$$y[n] = (-1)^n z[n] = (-1)^n \sum_{m=-\infty}^{\infty} (-1)^m x[m] h_L[n-m]$$

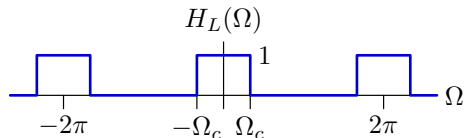
$$= \sum_{m=-\infty}^{\infty} (-1)^{n+m} x[m] h_L[n-m] = \sum_{m=-\infty}^{\infty} (-1)^{n-m} x[m] h_L[n-m]$$

$$= (x * h_H)[n] \quad \text{where} \quad h_H[n] = (-1)^n h_L[n]$$

## Cascaded System

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Find an expression for the unit sample response of a lowpass filter that passes frequencies in the range  $-\Omega_c < \Omega < \Omega_c$ .



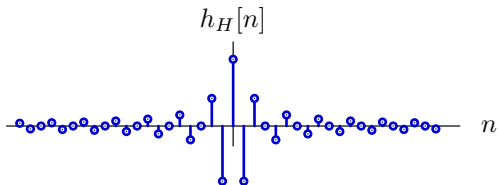
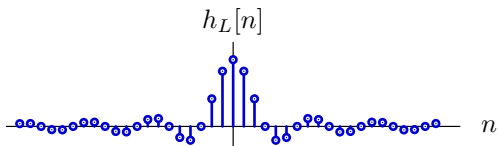
$$\begin{aligned}h_L[n] &= \frac{1}{2\pi} \int_{-\infty}^{\infty} H_L(\Omega) e^{j\Omega n} d\Omega \\&= \frac{1}{2\pi} \int_{-\Omega_c}^{\Omega_c} e^{j\Omega n} d\Omega \\&= \frac{1}{2\pi} \left[ \frac{e^{j\Omega n}}{jn} \right]_{-\Omega_c}^{\Omega_c} \\&= \frac{\sin(\Omega_c n)}{\pi n}\end{aligned}$$

$$\therefore h_H[n] = (-1)^n h_L[n] = (-1)^n \frac{\sin(\Omega_c n)}{\pi n}$$

## Cascaded System

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Plot  $h_L[n]$  and  $h_H[n]$  for  $\Omega_c = \frac{\pi}{3}$ .



## Cascaded System

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Determine an expression for  $H_H(\Omega)$ .

$$\begin{aligned}H_H(\Omega) &= \sum_{n=-\infty}^{\infty} h_H[n]e^{-j\Omega n} \\&= \sum_{n=-\infty}^{\infty} (-1)^n h_L[n]e^{-j\Omega n} \\&= \sum_{n=-\infty}^{\infty} e^{j\pi n} h_L[n]e^{-j\Omega n} \\&= \sum_{n=-\infty}^{\infty} h_L[n]e^{-j(\Omega-\pi)n} \\&= H_L(\Omega - \pi)\end{aligned}$$

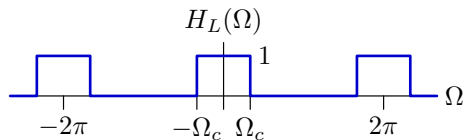
This is an example of the frequency shifting property:

$$\text{If } x[n] \xleftrightarrow{\text{DTFT}} X(\Omega) \text{ then } e^{j\Omega_o n} x[n] \xleftrightarrow{\text{DTFT}} X(\Omega - \Omega_o).$$

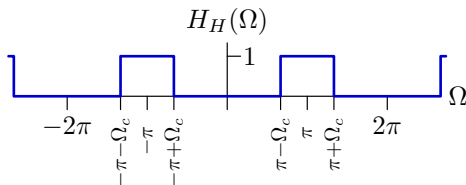
## Cascaded System

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Plot  $H_H(\Omega)$ . Compare to  $H_L(\Omega)$ .



$$H_H(\Omega) = H_L(\Omega - \pi)$$

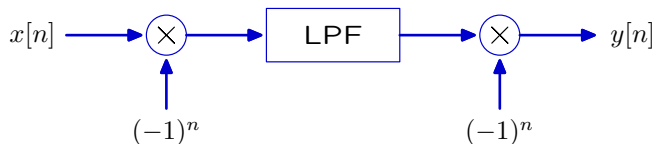


$H_H(\Omega)$  is a highpass filter!

## Cascaded System

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We have just shown that the original system is a highpass system.



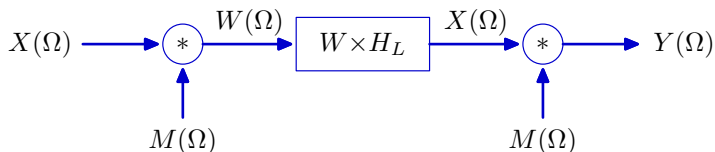
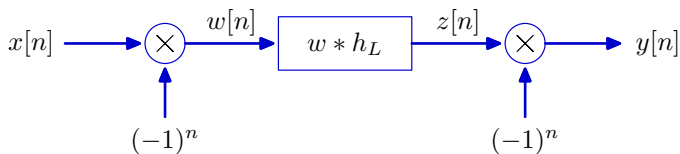
Therefore all of the following are true!

- The transformation from  $x[n]$  to  $y[n]$  is linear. ✓
- The transformation from  $x[n]$  to  $y[n]$  is time invariant. ✓
- The transformation from  $x[n]$  to  $y[n]$  is a high-pass filter. ✓

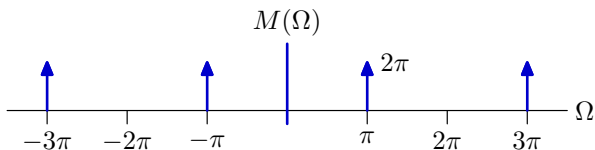


## Cascaded System

Alternatively, we could solve this problem in the frequency domain.



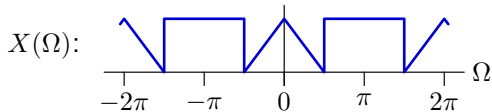
Since  $(-1)^n = e^{j\pi n}$ ,  $M(\Omega)$  is a complex sinusoid with frequency  $\pi$ .



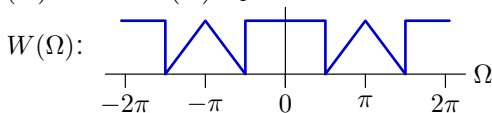
## Cascaded System

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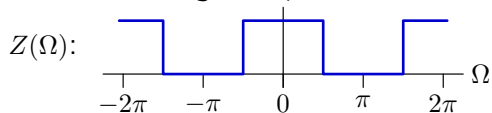
Assume  $X(\Omega)$  differs at high and low frequencies, as shown below.



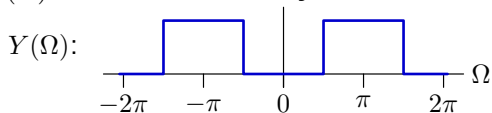
Convolving with  $M(\Omega)$  shifts  $X(\Omega)$  by  $\Omega = \pi$ :



Low pass filtering removes the high frequencies. Assume  $\Omega_c = \pi/2$ :



Convolving with  $M(\Omega)$  shifts the result by  $\Omega = \pi$ :



$Y(\Omega)$  is a highpass version of  $X(\Omega)$ .